

Synthesis from VHDL

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Outline

General assumptions

IEEE Standard logic package (Std_logic_1164)

Synthesis

- Combinational logic

- Sequential logic

- Finite State Machines (FSMs)

General assumptions



General assumptions

- only subset of VHDL is synthesizable,
- selected data types are supported,
- RTL and HLS synthesis (we concentrate mostly on RTL synthesis),
- description style has usually quite a big impact on the synthesized hardware,
- differences between different vendors.

IEEE Standard logic package (Std_logic_1164)



IEEE Standard logic package

- VHDL defines only basic features to model digital devices, such as, simple BIT type, two types of a delay, etc.
- more complex data types for signals must be defined for realistic simulation.

An Example:

```
entity nand_2 is
  port (I1, I2: in bit; O: out bit);
  -- interface description for
  -- two input nand gate
end nand_2;

architecture standard of nand_2 is
begin
  O <= NOT (I1 AND I2) after 2.5 ns;
end standard;
```

IEEE Standard logic package

Why we need more complex logic definition than BIT:

- we want to get more accurate simulation results,
- we use different technology and we want to simulate the same design in different technologies,
- real world is more complex than '0' and '1',
- more complex delays than standard VHDL models
 - inputs, output delays,
 - rising edge, falling edge delays,
 - min, max and typical delays,

Std_logic_1164

Requirements:

- support for more realistic signal modeling,
- technology independence, allowing extensions for the future and providing support for all current technologies,
- provide consistent features which facilitate logic level modeling,
- hide the complexity of the package from the designer as much as possible, make models readable,
- provide for the timing accuracy continuum, giving the modeler flexibility to choose the appropriate level of model accuracy.

Intended use of the packages:

```
-- access standard logic facilities
use ieee.Std_logic_1164.all;
-- VHDL entity declaration
--
```

Std_logic_1164

Requirements:

- modeling signals with '0' and '1' simplifies the real circuit, because it only consider signal voltage without signals current,
- many simulators introduce, so called, signal strength related to the signal's current,
- in std_logic_1164 we will introduce the following strengths of signals:
 - unknown,
 - forced (connected directly to ground or Vcc),
 - weak (connected to ground or Vcc through high resistive connection),
 - high impedance,
 - don't care.

Std_logic_1164

```
library IEEE;
PACKAGE Std_logic_1164 is
    -----
    -- Logic State System (unresolved)
    -----
    TYPE std_ulogic is ( 'U', -- Uninitialized
                        'X', -- Forcing Unknown
                        '0', -- Forcing 0
                        '1', -- Forcing 1
                        'Z', -- High Impedance
                        'W', -- Weak Unknown
                        'L', -- Weak 0
                        'H', -- Weak 1
                        '-' -- don't care
    );
```

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Std_logic_1164

```
-----
-- Unconstrained array of std_ulogic for use with
-- the resolution function
-----
TYPE std_ulogic_vector IS ARRAY
    ( NATURAL RANGE <> ) of std_ulogic;
-----
-- Resolution function
-----
FUNCTION resolved ( s : std_ulogic_vector )
    RETURN std_ulogic;
```

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Std_logic_1164

```
-----
-- *** Industry Standard Logic Type ***
-----
SUBTYPE std_logic IS resolved std_ulogic;
-----
-- Unconstrained array of std_logic for use in
-- declaring signal arrays
-----
TYPE std_logic_vector IS ARRAY
    ( NATURAL RANGE <> ) of std_logic;
```

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Std_logic_1164

```
-----  
-- Basic states + Test  
-----  
SUBTYPE X01 is resolved std_ulogic range 'X' to '1';  
-- ('X','0','1')  
SUBTYPE X01Z is resolved std_ulogic range 'X' to 'Z';  
-- ('X','0','1','Z')  
SUBTYPE UX01 is resolved std_ulogic range 'U' to '1';  
-- ('U','X','0','1')  
SUBTYPE UX01Z is resolved std_ulogic range 'U' to 'Z';  
-- ('U','X','0','1','Z')  
-----  
-- Overloaded Logical Operators  
-----  
FUNCTION "and" ( l : std_ulogic; r : std_ulogic )  
    RETURN UX01;  
:  
FUNCTION "not" ( l : std_ulogic )  
    RETURN UX01;
```

Use of Std_logic_1164

Example:

```
entity nand_2 is  
    port (I1, I2: in std_logic; O: out std_logic);  
    -- interface description for  
    -- two input nand gate using std_logic_1164  
end nand_2;
```

```
architecture standard of nand_2 is  
begin  
    O <= NOT (I1 AND I2) after 2.5 ns;  
end standard;
```

Synthesis



Synthesis

- different synthesis levels (behavioral, RTL, logic),
- different tools have different assumptions on specification styles,
- distinction between combinational logic and sequential logic.

Synthesis

Supported VHDL Language Constructs:

- **Entity**, **Architecture** and **Package** design units.
- **Function** and **Procedure** sub-programs
- IEEE Libraries - Std_Logic_1164, Std_Logic_Unsigned, Std_Logic_Signed, Numeric_Std and Numeric_Bit
- Ports of mode **in**, **out**, **inout** and **buffer**
- **Signals**, **Constants** and **Variables** (the latter should be restricted to sub-programs and processes)
- Composite types - **Arrays** and **Records**
- **Integer** and subtypes **Natural** and **Positive** (Integer types should have a range constraint unless a 32-bit word is required)

Synthesis (cont'd)

Supported VHDL Language Constructs:

- User defined enumeration types (eg. **type** State_type is (s0, s1, s2, s3);).
- Operators - +, -, *, /, **, **mod**, **rem**, **abs**, **not**, =, /=, <, >, <=, >=, **and**, **or**, **nand**, **nor**, **xor**, **xnor**, **sl**, **srl**, **sla**, **sra**, **rol**, **ror**, & . (Notes: /, mod and rem are usually supported for compile-time constants or when the right-hand argument is a power of 2. The shifting operators are usually supported for compile-time constant shift values)
- Sequential statements - signal and variable assignments, **wait**, **if**, **case**, **loop**, **for**, **while**, **return**, **null**, function and procedure call.

Note:

- only a single wait statement is allowed in a process,
- Only bounded loops are accepted.

Synthesis (cont'd)

Supported VHDL Language Constructs:

- Concurrent statements - signal assignment, **process**, **block**, component instantiation, sub-program call, **generate**.
- **Generic** ports in entities.
- Predefined attributes - 'range, 'event
- Aggregates and others clause.

Synthesis (cont'd)

Unsupported VHDL Language Constructs:

- **Access** and **File** types - the former are similar to C's pointers and files have no direct correspondence to hardware.
- **Register** and **Bus** kind signals - very rarely used VHDL constructs.
- Guarded blocks - as above, rarely used.
- **Next** and **Exit** loop control statements - A synthesis tool creates logic from a loop by 'unrolling' the loop into a series of computations, often resulting in iterative circuits. Prematurely terminating a loop prevents this unrolling process.

Synthesis (cont'd)

Unsupported VHDL Language Constructs:

- Objects of type Real - floating point numbers cannot be mapped to hardware and therefore are not supported.
- User defined resolution functions - prior to IEEE Standard 1164, designers made up their own multi-valued logic systems and resolution functions to support technology related aspects of simulation. None of these custom solutions is standard, and therefore none are supported by synthesis.

Synthesis (cont'd)

Ignored Constructs:

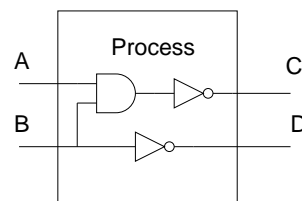
- **Assert** and **report** statements - these are included in a design to send messages to the 'console' window relaying information about what is happening during a simulation. As such, they have nothing to do with the hardware of the design.
- **After** clause - this is used to specify inertial and transport hardware delays in a design, or alternatively in a test-bench to produce clocks and other control waveforms. Synthesis tools have no way of creating a specific delay time, unless it is created by means of counting clock pulses in hardware. Delays may be included in the pre-synthesis RTL design. However, they will be ignored during the synthesis process.

Combinational process

```
comb_process:process(A, B)
begin
  C <= not(A and B) after 20 ns;
  D <= not B after 20 ns;
end process comb_process;
```

Difference with

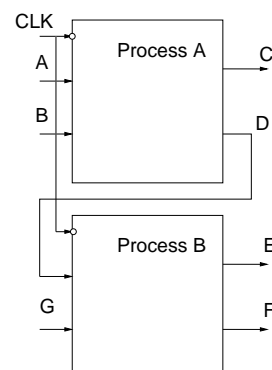
```
comb_process:process(A)
begin
  C <= not(A and B) after 20 ns;
  D <= not B after 20 ns;
end process comb_process;
```



Sequential process

```
A_process:process
begin
  wait clk'event and clk = '1'
  C <= not(A and B);
  D <= not B after 10 ns;
end process A_process;
```

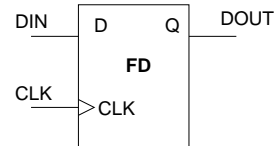
```
B_process:process
begin
  wait clk'event and clk = '1'
  E <= not(D and G);
  F <= not G;
end process B_process;
```



Sequential process

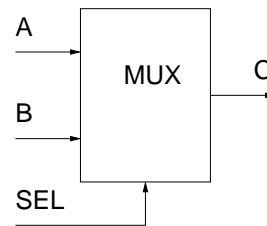
Clocked processes lead to all signals, assigned inside the process, in a flip-flop.

```
example:process
begin
    wait clk'event and clk = '1'
    dout <= din;
end process example;
```



If statement

```
if sel = '1'
    c <= b
else
    c <= a;
end if;
```



Asynchronous reset

```
process(clk, reset)
begin
    if reset = '1' then
        data <= '00';
    elseif clk'event and clk = '1' then
        data <= in_data;
    end if;
end process;
```

Synchronous reset

```
process(clk)
begin
  if clk'event and clk='1' then
    if reset = '1' then
      data <= "'00'";
    else
      data <= in_data;
    end if;
  end if;
end process;
```

Finite State Machines (FSMs)

- exist two types of FSMs, Moore and Mealy,
- contain register(s) for keeping FSM state,
- have logic for computing next state and output,
- can be modeled with combinational and sequential processes.

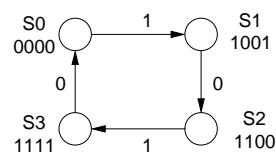
Moore machine

```
entity demo is
  port(clk, in1, reset: in std_logic;
        out1: out std_logic_vector(3 downto 0));
end demo;

architecture moore of demo is
  type state_type is (s0, s1, s2, s3);
  signal state: state_type;

begin
  state_process: process(clk, reset)
  :
  end process;

  output_process: process(state)
  :
  end process;
end moore;
```



Moore machine (cont'd)

```
state_process: process(clk, reset)
begin
  if reset = '1' then
    state <= s0;
  elseif clk'event and clk = '1' then
    case state is
      when s0 => if in1 = '1' then
        state <= s1; end if;
      when s1 => if in1 = '0' then
        state <= s2; end if;
      when s2 => if in1 = '1' then
        state <= s3; end if;
      when s3 => if in1 = '0' then
        state <= s0; end if;
    end case;
  end if
end process;
```

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Moore machine (cont'd)

```
output_process: process(state)
begin
  case state is
    when s0 => out1 <= "0000";
    when s1 => out1 <= "1001";
    when s2 => out1 <= "1100";
    when s3 => out1 <= "1111";
  end case;
end process;
```

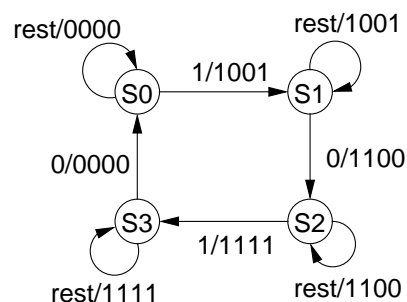
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Mealy machine



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Mealy machine (cont'd)

- state process – the same as for Moore machine
- output process

```
output_process: process(state)
begin
  case state is
    when s0 => if in1 = '1' then
                  out1 <= '1001';
                else
                  out1 <= '0000';
                end if;
    :
  end process;
```

Incompletely defined combinational processes

```
architecture bad is
begin
  process(state)
  begin
    if a > b then
      q <= '0';
    elseif a < b then
      q <= '1';
    end if;
  end process
end;
```

```
architecture good is
begin
  process(state)
  begin
    if a > b then
      q <= '0';
    else
      q <= '1';
    end if;
  end process
end;
```

Bad synthesis

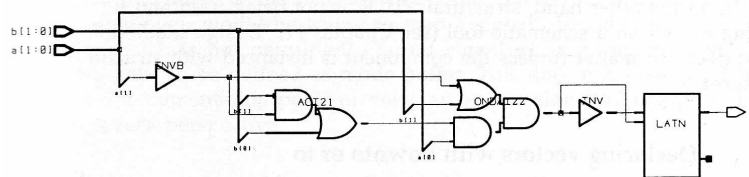


Figure 14.2: Synthesis result of an incomplete process.

Good synthesis

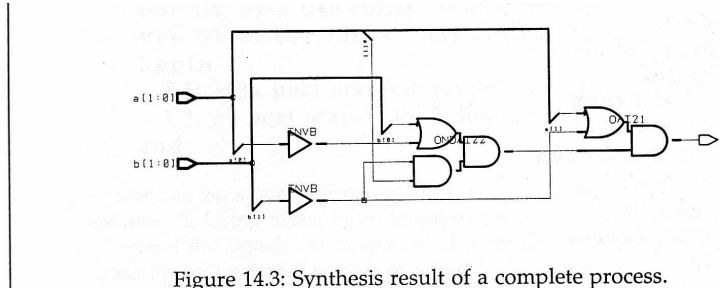


Figure 14.3: Synthesis result of a complete process.

High-level synthesis

High-level synthesis (behavioral, algorithm, architectural synthesis) means going from an algorithm specification to an RT level which implements the behavior.

- variables are allocated into registers or memory elements (sharing of registers is possible)- resource allocation,
- operators are allocated to functional units (several operators can be implemented by one functional unit, for example ALU)- resource allocation,
- operations are assigned to time slots for their execution in a synchronous implementation- scheduling.
- additional register are added if operations take more than one clock cycle.

High-level synthesis – example

Synthesis of the following code (inner loop of differential equation integrator)

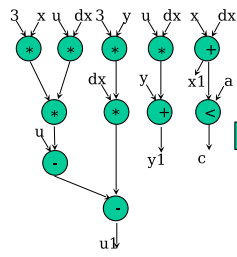
```

while c do
  begin
    x1 := x + dx;
    u1 := u - (3*x*u*dx);
    y1 := y + u*dx;
    c := x < a;
    x := x1;  y := y1;  u := u1;
  end;
end;

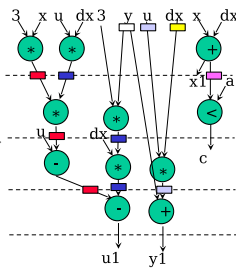
```

High-level synthesis – example

data-flow graph



scheduled
data-flow graph



register
allocation

