Outline

General assumptions

IEEE Standard logic package (Std_logic_1164)

Synthesis
  Combinational logic
  Sequential logic
  Finite State Machines (FSMs)

General assumptions
General assumptions

- only subset of VHDL is synthesizable,
- selected data types are supported,
- RTL and HLS synthesis (we concentrate mostly on RTL synthesis),
- description style has usually quite a big impact on the synthesized hardware,
- differences between different vendors.

IEEE Standard logic package (Std_logic_1164)

VHDL defines only basic features to model digital devices, such as, simple BIT type, two types of a delay, etc.

more complex data types for signals must be defined for realistic simulation.

An Example:

```
entity nand_2 is
  port (I1, I2: in bit; O: out bit);
  -- interface description for two input nand gate
end nand_2;

architecture standard of nand_2 is
begin
  O <= NOT (I1 AND I2) after 2.5 ns;
end standard;
```
IEEE Standard logic package

Why we need more complex logic definition than BiT:

- we want to get more accurate simulation results,
- we use different technology and we want to simulate the same design in different technologies,
- real world is more complex than '0' and '1',
- more complex delays than standard VHDL models
  - inputs, output delays,
  - rising edge, falling edge delays,
  - min, max and typical delays,

Std_logic_1164

Requirements:

- support for more realistic signal modeling,
- technology independence, allowing extensions for the future and providing support for all current technologies,
- provide consistent features which facilitate logic level modeling,
- hide the complexity of the package from the designer as much as possible, make models readable,
- provide for the timing accuracy continuum, giving the modeler flexibility to choose the appropriate level of model accuracy.

Intended use of the packages:

```vhdl
-- access standard logic facilities
use ieee.Std_logic_1164.all;
-- VHDL entity declaration
```

Std_logic_1164

Requirements:

- modeling signals with '0' and '1' simplifies the real circuit, because it only consider signal voltage without signals current,
- many simulators introduce, so called, signal strength related to the signal's current,
- in std_logic_1164 we will introduce the following strengths of signals:
  - unknown,
  - forced (connected directly to ground or Vcc),
  - weak (connected to ground or Vcc through high resistive connection),
  - high impedance,
  - don't care.
library IEEE;
PACKAGE Std_logic_1164 is
-------------------------------------------------
-- Logic State System (unresolved)
-------------------------------------------------
TYPE std_logic is ('U', -- Uninitialized
                  'X', -- Forcing Unknown
                  '0', -- Forcing 0
                  '1', -- Forcing 1
                  'Z', -- High Impedance
                  'W', -- Weak Unknown
                  'L', -- Weak 0
                  'H', -- Weak 1
                  '-' -- don't care);
-------------------------------------------------
-- Unconstrained array of std_logic for use with
-- the resolution function
-------------------------------------------------
TYPE std_logic_vector IS ARRAY (NATURAL RANGE <>) of std_logic;
-------------------------------------------------
-- Resolution function
-------------------------------------------------
FUNCTION resolved (s : std_logic_vector) RETURN std_logic;
-------------------------------------------------
-- *** Industry Standard Logic Type ***
-------------------------------------------------
SUBTYPE std_logic IS resolved std_logic;
-------------------------------------------------
-- Unconstrained array of std_logic for use in
-- declaring signal arrays
-------------------------------------------------
TYPE std_logic_vector IS ARRAY (NATURAL RANGE <>) of std_logic;
**Std_logic_1164**

---

-- Basic states + Test
---

SUBTYPE X01 is resolved std_ulogic range 'X' to '1';
-- ('X','0','1')

SUBTYPE X01Z is resolved std_ulogic range 'X' to 'Z';
-- ('X','0','1','Z')

SUBTYPE UX01 is resolved std_ulogic range 'U' to '1';
-- ('U','X','0','1')

SUBTYPE UX01Z is resolved std_ulogic range 'U' to 'Z';
-- ('U','X','0','1','Z')
---

-- Overloaded Logical Operators
---

FUNCTION "and" ( l : std_ulogic; r : std_ulogic )
RETURN UX01;

FUNCTION "not" ( l : std_ulogic )
RETURN UX01;

---

Kris Kuchcinski

**Use of Std_logic_1164**

Example:

entity nand_2 is
    port (I1, I2: in std_logic; O: out std_logic);
    -- interface description for
    -- two input nand gate using std_logic_1164
end nand_2;

architecture standard of nand_2 is
begin
    O <= NOT (I1 AND I2) after 2.5 ns;
end standard;

---

Kris Kuchcinski

**Synthesis**
Synthesis

- different synthesis levels (behavioral, RTL, logic),
- different tools have different assumptions on specification styles,
- distinction between combinational logic and sequential logic.

Synthesis

Supported VHDL Language Constructs:
- **Entity, Architecture** and **Package** design units.
- **Function** and **Procedure** sub-programs
- IEEE Libraries - **Std_Logic_1164**, **Std_Logic_Unsigned**, **Std_Logic_Signed**, **Numeric_Std** and **Numeric_Bit**
- Ports of mode **in**, **out**, **inout** and **buffer**
- **Signals**, **Constants** and **Variables** (the latter should be restricted to sub-programs and processes)
- Composite types - **Arrays** and **Records**
- **Integer** and subtypes **Natural** and **Positive** (Integer types should have a range constraint unless a 32-bit word is required)

Synthesis (cont’d)

Supported VHDL Language Constructs:
- User defined enumeration types (eg. **type State_type** is (s0, s1, s2, s3);).
- Operators - +, *, /, **, mod, rem, abs, not, =, /=, <, >, <=, >=, and, or, nand, nor, xor, xnor, sll, srl, sla, sra, rol, ror, &. (Notes: /, mod and rem are usually supported for compile-time constants or when the right-hand argument is a power of 2. The shifting operators are usually supported for compile-time constant shift values)
  Note:
  - only a single wait statement is allowed in a process,
  - Only bounded loops are accepted.
Synthesis (cont’d)

Supported VHDL Language Constructs:
- Concurrent statements - signal assignment, process, block, component instantiation, sub-program call, generate.
- Generic ports in entities.
- Predefined attributes - `range, `event
- Aggregates and others clause.

Synthesis (cont’d)

Unsupported VHDL Language Constructs:
- Access and File types - the former are similar to C’s pointers and files have no direct correspondence to hardware.
- Register and Bus kind signals - very rarely used VHDL constructs.
- Guarded blocks - as above, rarely used.
- Next and Exit loop control statements - A synthesis tool creates logic from a loop by ‘unrolling’ the loop into a series of computations, often resulting in iterative circuits. Prematurely terminating a loop prevents this unrolling process.

Synthesis (cont’d)

Unsupported VHDL Language Constructs:
- Objects of type Real - floating point numbers cannot be mapped to hardware and therefore are not supported.
- User defined resolution functions - prior to IEEE Standard 1164, designers made up their own multi-valued logic systems and resolution functions to support technology related aspects of simulation. None of these custom solutions is standard, and therefore none are supported by synthesis.
Synthesis (cont’d)

Ignored Constructs:

- **Assert and report** statements - these are included in a design to send messages to the ‘console’ window relaying information about what is happening during a simulation. As such, they have nothing to do with the hardware of the design.

- **After** clause - this is used to specify inertial and transport hardware delays in a design, or alternatively in a test-bench to produce clocks and other control waveforms. Synthesis tools have no way of creating a specific delay time, unless it is created by means of counting clock pulses in hardware. Delays may be included in the pre-synthesis RTL design. However, they will be ignored during the synthesis process.

Combinational process

```
comb_process: process(A, B)
begin
  C <= not(A and B) after 20 ns;
  D <= not B after 20 ns;
end process comb_process;
```

**Difference with**

```
comb_process: process(A)
begin
  C <= not(A and B) after 20 ns;
  D <= not B after 20 ns;
end process comb_process;
```

Sequential process

```
A_process: process
begin
  wait clk’event and clk = '1'
  C <= not(A and B);
  D <= not B after 10 ns;
end process A_process;
```

```
B_process: process
begin
  wait clk’event and clk = '1'
  E <= not(D and G);
  F <= not G;
end process B_process;
```
Sequential process

Clocked processes lead to all signals, assigned inside the process, in a flip-flop.

Example:

```vhdl
example: process
begin
    wait clk'event and clk = '1'
    dout <= din;
end process example;
```

If statement

```vhdl
if sel = '1'
    c <= b
else
    c <= a;
end if;
```

Asynchronous reset

```vhdl
process(clk, reset)
begin
    if reset = '1' then
        data <= '"00"';
    elseif clk'event and clk = '1' then
        data <= in_data;
    end if;
end process;
```
Synchronous reset

```vhdl
process(clk)
begin
  if clk'event and clk='1' then
    if reset = '1' then
      data <= "00";
    else
      data <= in_data;
    end if;
  end if;
end process;
```

Finite State Machines (FSMs)

- exist two types of FSMs, Moore and Mealy,
- contain register(s) for keeping FSM state,
- have logic for computing next state and output,
- can be modeled with combinational and sequential processes.

Moore machine

```vhdl
entity demo is
  port(clk, in1, reset: in std_logic;
       out1: out std_logic_vector(3 downto 0);
end demo;
architecture moore of demo is
  type state_type is (s0, s1, s2, s3);
  signal state: state_type;
begin
  state_process: process(clk, reset)
  begin
    state <= state + 1;
  end process;
  output_process: process(state)
  begin
    out1 <= state;
  end process;
end moore;
```
Moore machine (cont’d)

```
state_process: process(clk, reset)
begin
  if reset = '1' then
    state <= s0;
  elseif clk'event and clk = '1' then
    case state is
    when s0 => if in1 = '1' then
      state <= s1;
    end if;
    when s1 => if in1 = '0' then
      state <= s2;
    end if;
    when s2 => if in1 = '1' then
      state <= s3;
    end if;
    when s3 => if in1 = '0' then
      state <= s0;
    end if;
  end case;
end if
end process;
```

Kris Kuchcinski Synthesis from VHDL

Moore machine (cont’d)

```
output_process: process(state)
begin
  case state is
  when s0 => out1 <= '0000';
  when s1 => out1 <= '1001';
  when s2 => out1 <= '1100';
  when s3 => out1 <= '1111';
  end case;
end process;
```

Kris Kuchcinski Synthesis from VHDL

Mealy machine

```
S0 S1
S2 S3
1/1001
0/1100
1/1111
0/0000
rest/0000
rest/1001
rest/1100
1/1111
rest/1100
```

Kris Kuchcinski Synthesis from VHDL
Mealy machine (cont’d)

- state process – the same as for Moore machine
- output process

```vhdl
output_process: process(state)
begin
  case state is
    when s0 =>
      if in1 = '1' then
        out1 <= '1001';
      else
        out1 <= '0000';
      end if;
  end process;
```

Kris Kuchcinski
Synthesis from VHDL

Incompletely defined combinational processes

```vhdl
architecture bad is
begin
  process(state)
  begin
    if a > b then
      q <= '0';
    elseif a < b then
      q <= '1';
    end if;
  end process
end;
```

```vhdl
architecture good is
begin
  process(state)
  begin
    if a > b then
      q <= '0';
    else
      q <= '1';
    end if;
  end process
end;
```

Kris Kuchcinski
Synthesis from VHDL

Bad synthesis

![Figure 14.2: Synthesis result of an incomplete process.](image-url)
High-level synthesis

High-level synthesis (behavioral, algorithm, architectural synthesis) means going from an algorithm specification to an RT level which implements the behavior.

- variables are allocated into registers or memory elements (sharing of registers is possible)- resource allocation.
- operators are allocated to functional units (several operators can be implemented by one functional unit, for example ALU)- resource allocation.
- operations are assigned to time slots for their execution in a synchronous implementation- scheduling.
- additional register are added if operations take more than one clock cycle.

High-level synthesis – example

Synthesis of the following code (inner loop of differential equation integrator)

```vhdl
while c do
    begin
        x1 := x + dx;
        ul := u - (3*x+u+dx);
        y1 := y + u*dx;
        c := x < a;
        x := x1; y := y1; u := ul;
    end;
```

Good synthesis

![Synthesis result of a complete process.](image)
High-level synthesis – example

data-flow graph

scheduled data-flow graph

register allocation

Kris Kuchcinski
Synthesis from VHDL

36